

ENGI 7825: Control Systems II

The State-Space Representation: Part 1

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Adapted from the notes of
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What to Expect

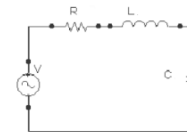
- ▶ Math-heavy, implementation-light
 - Background: Linear algebra, differential equations, stuff from 5821
- ▶ Assignments: Mostly paper and pen, some Matlab
- ▶ Practical Assignments: Using V-REP robot simulator
 - The purpose is to motivate the techniques discussed and illustrate how they might be useful. Example problems drawn from mobile robotics.
- ▶ Mid-term tests and final: Based on core material presented in class and practiced on the assignments
- ▶ Note: This course is not a hands-on introduction to industrial controls. It focuses instead on control systems theory.
 - (But remember that theory is the stuff that is harder to learn after graduating)

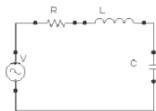
Why a Second Controls Course?

- ▶ Classical control techniques (ENGI 5821) only applicable to single-input, single-output (SISO) systems:
 - A heating system for one room
 - Orientation of a single joint on a robot
- ▶ Many real-world systems have multiple inputs and multiple outputs (MIMO)
 - A heating system for two rooms that accounts for heat transfer between them:
 - Input/Output: 2 temperatures
 - Simple two-wheeled mobile robot:
 - Input: 2 wheel speeds
 - Output: translational and rotational velocity
- ▶ Also, classical control techniques are usually based on the assumption of zero initial conditions; Lets review...

Classical Control

- ▶ The classical (i.e. frequency-domain) approach: converting a system's differential equation to a **transfer function**, thus generating a mathematical model of the system that relates the input to the output
- ▶ e.g. RLC circuit with $v(t)$ as input and $v_c(t)$ as output





input: $v(t)$
output: $v_c(t)$

KVL: $Ri(t) + L \frac{di(t)}{dt} + v_c(t) = v(t)$

We need to relate the input $v(t)$ to the output, $v_c(t)$; Replace $i(t)$:

$$i(t) = C \frac{dv_c(t)}{dt}$$

$$RC \frac{dv_c(t)}{dt} + LC \frac{d^2v_c(t)}{dt^2} + v_c(t) = v(t)$$

$$LC \frac{d^2v_c(t)}{dt^2} + RC \frac{dv_c(t)}{dt} + v_c(t) = v(t)$$

$$LC \frac{d^2v_c(t)}{dt^2} + RC \frac{dv_c(t)}{dt} + v_c(t) = v(t)$$

► Apply Laplace Transform using the following theorems:

Theorem

$$\mathcal{L}\left\{\frac{df}{dt}\right\} = sF(s) - f(0-)$$

$$\mathcal{L}\left\{\frac{d^2f}{dt^2}\right\} = s^2F(s) - sf(0-) - f'(0-)$$

$$\mathcal{L}\left\{\frac{d^nf}{dt^n}\right\} = s^nF(s) - \sum_{k=1}^n s^{n-k} f^{(k-1)}(0-)$$

► If the initial conditions are all zero then we get

$$LCs^2V_c(s) + RCsV_c(s) + V_c(s) = V(s)$$

► If the initial conditions are non-zero we get

$$LCs^2V_c(s) + RCsV_c(s) + V_c(s) - LCsv_c(0) - LCv'_c(0) - RCv_c(0) = V(s)$$

► Zero initial conditions:

$$LCs^2V_c(s) + RCsV_c(s) + V_c(s) = V(s)$$



$$\frac{V_c(s)}{V(s)} = \frac{1/LC}{s^2 + \frac{R}{L}s + \frac{1}{LC}}$$

► This is a transfer function (output / input)

► With non-zero initial conditions, there is no way to get a transfer function:

$$LCs^2V_c(s) + RCsV_c(s) + V_c(s) - LCsv_c(0) - LCv'_c(0) - RCv_c(0) = V(s)$$



$$\frac{V_c(s)}{V(s)} = ?$$

Advantages of State-Space Controls

► State-Space Controls is a method for modeling, analyzing and designing a wide range of systems, such as:

- Nonzero initial condition systems
 - Classical restricted to zero initial conditions
- Multiple-input, multiple-output systems (MIMO)
 - Classical restricted to single-input, single-output system (SISO)

► State-space has other advantages that we will see later on. For example:

- Natural representation for nonlinear systems, time-varying systems (variable mass systems, ...)

System State

- ▶ The state of a system is the information needed in addition to the input to determine the output
- ▶ A **memoryless** system's output is determined solely by the input
 - e.g. A single resistor with voltage as input and current as output
- ▶ The output of a **causal** system depends on past and current inputs. The impact of past inputs is represented by the state.
 - All physically realized systems are causal
- ▶ The output of a **non-causal** system depends on past, current, and future inputs! No real-world physical systems have this ability to anticipate future inputs
 - Non-causal systems can be simulated and non-causal filtering can be applied on previously collected data

Notation

- ▶ We would like to describe general MIMO systems, so the inputs, outputs, and states will all be vectors
- ▶ input vector: $\mathbf{u}(t)$
 - Assume we have control over the input
- ▶ output vector: $\mathbf{y}(t)$
 - The output is any single variable or multiple variables that we can measure
- ▶ state vector: $\mathbf{x}(t)$
 - Other variables representing the state of the system (may or may not be measurable)

State-Space Representation

- ▶ The state-space representation for an LTI system has the general form:

$$\begin{cases} \dot{\mathbf{x}}(t) = \mathbf{A}\mathbf{x}(t) + \mathbf{B}\mathbf{u}(t) \\ \mathbf{y}(t) = \mathbf{C}\mathbf{x}(t) + \mathbf{D}\mathbf{u}(t) \end{cases} \quad \mathbf{x}(t_0) = \mathbf{x}_0$$

Initial conditions

where:

$\mathbf{x}(t)$: state vector (n-dimensional)

$\mathbf{y}(t)$: output vector (p-dim)

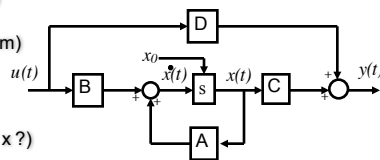
$\mathbf{u}(t)$: input or control vector (m-dim)

A: dynamic matrix (? x ?)

B: input matrix (? x ?)

C: output matrix (? x ?)

D: feedforward (direct) matrix (? x ?)



Definitions

- ▶ **System state**: minimum information needed in order to completely determine the output of a system from a given moment, provided the input is known from that moment
- ▶ **System variable**: any variable that responds to an input or initial conditions in a system
- ▶ **State variables**: the smallest set of linearly independent system variables such that the values of the set members at time t_0 along with known inputs completely determine the value of all system variables for all $t \geq t_0$
- ▶ **State vector**: vector whose elements are the state variables
- ▶ **State space**: n-dimensional space whose axes represent the state variables
- ▶ **State equations**: set of n simultaneous, first order differential equations with n variables, where the n variables to be solved are the state variables
- ▶ **Output equation**: algebraic expression of the output variables of a system as linear combinations of the state variables and the inputs

Procedure for Selecting State Variables

$$\begin{aligned} \dot{\mathbf{x}}(t) &= \mathbf{Ax}(t) + \mathbf{Bu}(t) \\ \mathbf{y}(t) &= \mathbf{Cx}(t) + \mathbf{Du}(t) \quad \mathbf{x}(t_0) = \mathbf{x}_0 \end{aligned}$$

- Consider the energy-storage elements in the system (capacitors and inductors in electrical syst.; masses and springs in mechanical syst.)
- For electrical systems you would then write the derivative equation for each energy-storage element

	Voltage-current	Current-voltage
Resistor	$v(t) = Ri(t)$	$i(t) = \frac{1}{R}v(t)$
Inductor	$v(t) = L \frac{di(t)}{dt}$	$i(t) = \frac{1}{L} \int_0^t v(\tau) d\tau$
Capacit.	$v(t) = \frac{1}{C} \int_0^t i(\tau) d\tau$	$i(t) = C \frac{dv(t)}{dt}$

- So for inductors, choose current as the state variable, for capacitors choose voltage
- Another perspective is to consider the relevant quantities in the energy equations: (inductor $E = \frac{1}{2} L i^2$, capacitor: $E = \frac{1}{2} C v^2$)

What do you do after selecting the state variables?

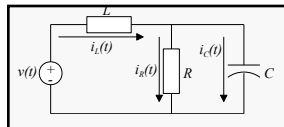
- Solve for the derivative of each state variable as a linear combination of system variables and the input
- Stack these scalar equations to form the state and output equations which operate on vectors
- The following is for a system with $n = 4$, $p = 2$ (outputs), $m = 2$ (inputs)

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \\ \dot{x}_4 \end{bmatrix} = \begin{bmatrix} ? & ? & ? & ? \\ ? & ? & ? & ? \\ ? & ? & ? & ? \\ ? & ? & ? & ? \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} + \begin{bmatrix} ? \\ ? \\ ? \\ ? \end{bmatrix} u(t) \quad \begin{bmatrix} y_1 \\ y_2 \end{bmatrix} = \begin{bmatrix} ? & ? & ? & ? \\ ? & ? & ? & ? \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} + \begin{bmatrix} ? \\ ? \end{bmatrix} u(t)$$

$$\begin{aligned} \dot{\mathbf{x}}(t) &= \mathbf{Ax}(t) + \mathbf{Bu}(t) \\ \mathbf{y}(t) &= \mathbf{Cx}(t) + \mathbf{Du}(t) \end{aligned}$$

State-space introductory example

- Example 1: Find a state-space representation of the system shown in the figure if the output is the current through the resistor.



- **Step 1** Select the state variables: write the derivative equation for the energy-storage elements (L and C)...

$$C \frac{dv_C}{dt} = i_C \quad L \frac{di_L}{dt} = v_L$$

and choose the differentiated quantities as the state variables (v_C , i_L)

$$C \frac{dv_C}{dt} = i_C \quad L \frac{di_L}{dt} = v_L$$

- **Step 2** Write the right-hand sides of "step 1" and the output equations as linear combinations of the state variables and the input.

$$\begin{aligned} C \frac{dv_C}{dt} = i_C &= f_1(v_C, i_L, v(t)) & i_C &= -i_R + i_L = -\frac{1}{R}v_C + i_L \\ L \frac{di_L}{dt} = v_L &= f_2(v_C, i_L, v(t)) & v_L &= -v_C + v(t) \end{aligned}$$

$$\text{State eq} \quad \begin{aligned} \frac{dv_C}{dt} &= -\frac{1}{RC}v_C + \frac{1}{L}i_L \\ \frac{di_L}{dt} &= -\frac{1}{L}v_C + \frac{1}{L}v(t) \end{aligned}$$

$$\text{Output eq} \quad i_R = f_3(v_C, i_L, v(t)) \quad i_R = \frac{1}{R}v_C$$

$$\frac{dv_c}{dt} = -\frac{1}{RC}v_c + \frac{1}{C}i_L$$

$$\frac{di_L}{dt} = -\frac{1}{L}v_c + \frac{1}{L}v(t)$$

$$i_R = \frac{1}{R}v_c$$

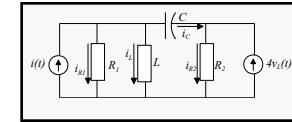
- Step 4 Obtain the S-S representation in vector-matrix form

$$\begin{bmatrix} \dot{v}_c \\ \dot{i}_L \end{bmatrix} = \begin{bmatrix} -\frac{1}{RC} & \frac{1}{C} \\ -\frac{1}{L} & 0 \end{bmatrix} \begin{bmatrix} v_c \\ i_L \end{bmatrix} + \begin{bmatrix} 0 \\ \frac{1}{L} \end{bmatrix} v(t)$$

$$i_R = \begin{bmatrix} \frac{1}{R} & 0 \end{bmatrix} \begin{bmatrix} v_c \\ i_L \end{bmatrix}$$

Another electrical example

- Example 2: Obtain the SS equations for the next circuit, with input $i(t)$ and output vector $y = [v_{R2} \ i_{R2}]^T$



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$$\text{States Eq} \quad \begin{bmatrix} \dot{v}_c \\ \dot{i}_L \end{bmatrix} = \begin{bmatrix} \frac{1}{R_1 C \Delta} & \frac{1-4R_2}{C \Delta} \\ -\frac{1}{L \Delta} & \frac{R_2}{L \Delta} \end{bmatrix} \begin{bmatrix} v_c \\ i_L \end{bmatrix} + \begin{bmatrix} -\frac{1-4R_2}{C \Delta} \\ -\frac{R_2}{L \Delta} \end{bmatrix} i(t)$$

$$\text{Out Eq} \quad \begin{bmatrix} v_{R2} \\ i_{R2} \end{bmatrix} = \begin{bmatrix} -(1 + \frac{1}{\Delta}) & \frac{R_2}{\Delta} \\ -\frac{(1+\frac{1}{\Delta})}{R_2} & \frac{1}{\Delta} \end{bmatrix} \begin{bmatrix} v_c \\ i_L \end{bmatrix} + \begin{bmatrix} -\frac{R_2}{\Delta} \\ -\frac{1}{\Delta} \end{bmatrix} i(t)$$

$$\text{Where} \quad \Delta = -(1 - 4R_2) + \frac{R_2}{R_1}$$

Mechanical systems

- The procedure for mechanical systems is a bit different, but we will still assign a state variable for each energy-storage element

TABLE 2.4 Force-velocity, force-displacement, and impedance translational relationships for springs, viscous dampers, and mass

Component	Force-velocity	Force-displacement	Impedance $Z_M(s) = F(s)/X(s)$
Spring 	$f(t) = K \int_0^t v(\tau) d\tau$	$f(t) = Kx(t)$	K
Viscous damper 	$f(t) = f_v v(t)$	$f(t) = f_v \frac{dx(t)}{dt}$	$f_v s$
Mass 	$f(t) = M \frac{dv(t)}{dt}$	$f(t) = M \frac{d^2x(t)}{dt^2}$	Ms^2

Energy storage elements

Note: The following set of symbols and units is used throughout this book: $f(t) = N$ (newtons), $x(t) = m$ (meters), $v(t) = m/s$ (meters/second), $K = N/m$ (newtons/meter), $f_v = N \cdot s/m$ (newton-seconds/meter), $M = kg$ (kilograms = newton-seconds²/meter).

- A spring has potential energy and a moving mass has kinetic energy. A viscous damper is analogous to a resistor in that it does not store energy.

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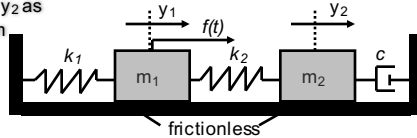
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- A spring has potential energy $\frac{1}{2} k y(t)^2$ so we will use displacement, $y(t)$ as a state variable
- A moving mass has kinetic energy $\frac{1}{2} m \dot{y}(t)^2$ so we will use the speed, $\dot{y}(t)$, as a state variable

Shows $x(t)$ as displacement but we'll use $y(t)$ since that is the intended output in our examples

A mechanical example

- Example 2: Find the state eqs for the translational mechanical system shown in figure, with y_1 and y_2 as outputs of the system and $f(t)$ the input

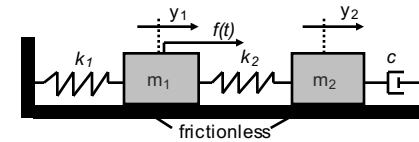


- What state variables should we define?
- For each mass we use the displacement (for the spring's potential energy) and speed (for the mass's kinetic energy)
- The following are all reasonable choices for state variables:

$$x_1(t) = y_1(t) \quad x_2(t) = y_2(t) \quad x_3(t) = \dot{y}_1(t) \quad x_4(t) = \dot{y}_2(t)$$

$$x_1(t) = y_1(t) \quad x_2(t) = \dot{y}_1(t) \quad x_3(t) = y_2(t) \quad x_4(t) = \dot{y}_2(t)$$

$$x_1(t) = y_1(t) \quad x_2(t) = y_2(t) - y_1(t) \quad x_3(t) = \dot{y}_1(t) \quad x_4(t) = \dot{y}_2(t)$$



We use the following state variables:

$$x_1(t) = y_1(t) \quad x_2(t) = y_2(t) \quad x_3(t) = \dot{y}_1(t) \quad x_4(t) = \dot{y}_2(t)$$

Consider the motion of each mass and apply Newton's second law: $\mathbf{m} \mathbf{a} = \Sigma \mathbf{F}$.

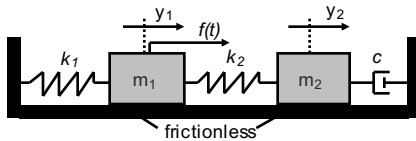
First consider the forces on m_1 :

$$m_1 \ddot{y}_1(t) = -k_1 y_1(t) - k_2 y_1(t) + k_2 y_2(t) + f(t)$$

$$m_1 \ddot{y}_1(t) = -(k_1 + k_2) y_1(t) + k_2 y_2(t) + f(t)$$

Now consider the forces on m_2 :

$$m_2 \ddot{y}_2(t) = k_2 y_1(t) - k_2 y_2(t) - c \dot{y}_2(t)$$



$$x_1(t) = y_1(t) \quad x_2(t) = y_2(t) \quad x_3(t) = \dot{y}_1(t) \quad x_4(t) = \dot{y}_2(t)$$

We derived the following on the previous slide:

$$m_1 \ddot{y}_1(t) = -(k_1 + k_2) y_1(t) + k_2 y_2(t) + f(t)$$

$$m_2 \ddot{y}_2(t) = k_2 y_1(t) - k_2 y_2(t) - c \dot{y}_2(t)$$

Combining into matrix-vector form yields the following SS representation:

$$\begin{bmatrix} \dot{x}_1 \\ \dot{x}_2 \\ \dot{x}_3 \\ \dot{x}_4 \end{bmatrix} = \begin{bmatrix} 0 & 0 & 1 & 0 \\ 0 & 0 & 0 & 1 \\ -\frac{k_1+k_2}{m_1} & \frac{k_2}{m_1} & 0 & 0 \\ \frac{k_2}{m_2} & -\frac{k_2}{m_2} & 0 & -\frac{c}{m_2} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \\ \frac{1}{m_1} \\ 0 \end{bmatrix} f(t) \quad \begin{bmatrix} y_1 \\ y_2 \end{bmatrix} = \begin{bmatrix} 1 & 0 & 0 & 0 \\ 0 & 1 & 0 & 0 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ x_3 \\ x_4 \end{bmatrix} + \begin{bmatrix} 0 \\ 0 \end{bmatrix} f(t)$$